







LMK1D1204, LMK1D1208 SNAS815A - DECEMBER 2020 - REVISED AUGUST 2021

LMK1D120x Low Additive Jitter LVDS Buffer

1 Features

- High-performance LVDS clock buffer family with 2 inputs and 4 (2:4) or 8 (2:8) outputs.
- Output frequency up to 2 GHz.
- Supply voltage: 1.71 V to 3.465 V
- Low additive jitter: < maximum 60 fs RMS in 12kHz to 20-MHz at 156.25 MHz
 - Very low phase noise floor: –164 dBc/Hz (typical)
- Very low propagation delay: < 575 ps maximum
- Output skew: 20 ps maximum
- Universal inputs accept LVDS, LVPECL, LVCMOS, LP-HCSL, HCSL and CML inputs
- LVDS reference voltage, V_{AC REF}, available for capacitive-coupled inputs
- Industrial temperature range: -40°C to 105°C
- Packages available:
 - LMK1D1204: 3-mm × 3-mm, 16-pin VQFN
 - LMK1D1208: 5-mm × 5-mm, 28-pin VQFN (RHD)

2 Applications

- Telecommunications and networking
- Medical imaging
- Test and measurement
- Wireless infrastructure
- Pro audio, video and signage

3 Description

The LMK1D120x clock buffer distributes one of two selectable clock inputs (IN0 and IN1) to 4 or 8 pairs of differential LVDS clock outputs (OUT0 through OUT7) with minimum skew for clock distribution. The LMK1D12x family can accept two clock sources into an input multiplexer. The inputs can either be LVDS, LVPECL, LP-HCSL, HCSL, CML or LVCMOS.

The LMK1D12x is specifically designed for driving 50- Ω transmission lines. In case of driving the inputs in single-ended mode, the appropriate bias voltage as shown in Figure 8-6 must be applied to the unused negative input pin.

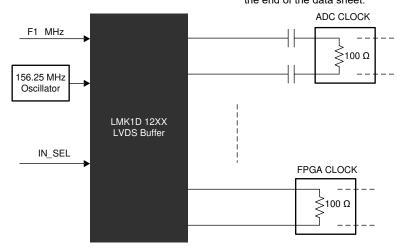
The IN SEL pin selects the input which is routed to the outputs. If this pin is left open, it disables the outputs (logic low). The part supports a fail-safe function. The device further incorporates an input hysteresis which prevents random oscillation of the outputs in the absence of an input signal.

The device operates in 1.8-V or 2.5-V or 3.3-V supply environment and is characterized from -40°C to 105°C (ambient temperature). The LMK1D12x package variant is shown in the table below:

Device Information

PART NUMBER ⁽¹⁾	PACKAGE	BODY SIZE (NOM)
LMK1D1204	VQFN (16)	3.00 mm × 3.00 mm
LMK1D1208	VQFN (28)	5.00 mm × 5.00 mm

For all available packages, see the orderable addendum at the end of the data sheet.



Application Example



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4 Revision History

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

CI	hanges from Revision * (December 2020) to Revision A (August 2021)	Page
•	First public release	1



5 Pin Configuration and Functions

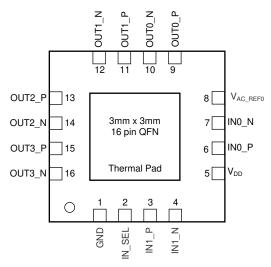


Figure 5-1. LMK1D1204: RGT Package 16-Pin VQFN Top View

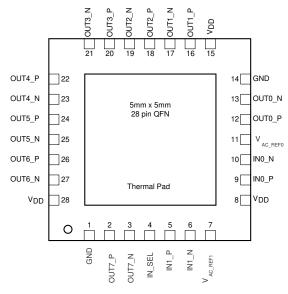


Figure 5-2. LMK1D1208: RHD Package 28-Pin VQFN Top View

Table 5-1. Pin Functions

	PIN		PIN		TYPE ⁽¹⁾	DESCRIPTION	
NAME	LMK1D1204	LMK1D1208	ITPE\/	DESCRIPTION			
DIFFERENTIAL/SINGLE	-ENDED CLOCK	INPUT					
IN0_P	6	9		Primary: Differential input pair or single-ended input			
INO_N	7	10	'	Filmary. Differential input pair or single-ended input			
IN1_P	3	5		Secondary: Differential input pair or single-ended input.			
IN1_N	4	6		Note that INP0, INN0 are used indistinguishably with IN0_P, IN0_N.			
INPUT SELECT							
IN_SEL	_SEL 2 4 I Input Selection with an internal 500-kΩ pullup and 320-kΩ pulldown resistor, selects input port; (See Table 8-1)		Input Selection with an internal 500-kΩ pullup and 320-kΩ pulldown resistor, selects input port; (See Table 8-1)				
BIAS VOLTAGE OUTPU	IAS VOLTAGE OUTPUT						



Table 5-1. Pin Functions (continued)

SCRIPTION		
used, TI		
his pin.		
for heat		
f		

⁽¹⁾ G = Ground, I = Input, O = Output, P = Power

6 Specifications

6.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted)(1)

		MIN	MAX	UNIT
V_{DD}	Supply voltage	-0.3	3.6	V
V _{IN}	Input voltage	-0.3	3.6	V
Vo	Output voltage	-0.3	V _{DD} + 0.3	V
I _{IN}	Input current	-20	20	mA
Io	Continuous output current	-50	50	mA
T _J	Junction temperature		135	°C
T _{stg}	Storage temperature (2)	-65	150	°C

⁽¹⁾ Stresses beyond those listed under *Absolute Maximum Rating* may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under *Recommended Operating Condition*. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

6.2 ESD Ratings

			VALUE	UNIT
V	Floatroatatic disphares	Human body model (HBM), per ANSI/ESDA/ JEDEC JS-001, all pins ⁽¹⁾	±3000	V
V _(ESD)	Electrostatic discharge	Charged device model (CDM), per JEDEC specification JESD22-C101, all pins ⁽²⁾	±1000	V

⁽¹⁾ JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.

6.3 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)

		·	MIN	NOM	MAX	UNIT
		3.3-V supply	3.135	3.3	3.465	
V_{DD}	Core supply voltage	2.5-V supply	2.375	2.5	2.625	V
		1.8-V supply	1.71	1.8	1.89	
Supply Ramp	Supply voltage ramp	Requires monotonic ramp (10-90% of V_{DD})	0.1		20	ms
T _A	Operating free-air temperature		-40		105	°C
T _J	Operating junction temperature		-40		135	°C

⁽²⁾ Device unpowered

⁽²⁾ JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.



6.4 Thermal Information

		LMK1D1204	LMK1D1208	
	THERMAL METRIC(1)	VQFN	VQFN	UNIT
		16 PINS	28 PINS	
$R_{\theta JA}$	Junction-to-ambient thermal resistance	48.7	38.9	°C/W
R _{0JC(top)}	Junction-to-case (top) thermal resistance	56.4	32.1	°C/W
R _{0JB}	Junction-to-board thermal resistance	23.6	18.7	°C/W
Ψ_{JT}	Junction-to-top characterization parameter	1.6	1	°C/W
Ψ_{JB}	Junction-to-board characterization parameter	23.6	18.7	°C/W
$R_{\theta JC(bot)}$	Junction-to-case (bottom) thermal resistance	8.6	8.2	°C/W

⁽¹⁾ For more information about traditional and new thermal metrics, see the Semiconductor and IC Package Thermal Metrics application report.

6.5 Electrical Characteristics

VDD = 1.8 V \pm 5 %, -40 °C \leq T A \leq 105 °C. Typical values are at VDD = 1.8 V, 25 °C (unless otherwise noted)

UNIT	MAX	MIN TYP	TEST CONDITIONS	PARAMETER	
				PPLY CHARACTERISTICS	POWER SUP
mA		50	All-outputs enabled and unterminated, f = 0 Hz	LMK1D1204	IDD _{STAT}
mA		55	All-outputs enabled and unterminated, f = 0 Hz	LMK1D1208	IDD _{STAT}
mA	72	60	All-outputs enabled, R_L = 100 Ω , f =100 MHz	LMK1D1204	IDD _{100M}
mA	95	78	All-outputs enabled, $R_L = 100 \Omega$, f = 100 MHz	LMK1D1208	IDD _{100M}
		d 3.3 V ± 5%)	ies to V _{DD} = 1.8 V ± 5%, 2.5 V ± 5%	ITROL INPUT CHARACTERISTICS (Appl	IN_SEL CON
V		0.4*V _{CC}	Open	3-state input	Vd _{I3}
V	V _{CC} + 0.3	0.7*V _{CC}	Minimun input voltage for a logical "1" state in table 1	Input high voltage	V _{IH}
V	0.3*V _{CC}	-0.3	Maximum input voltage for a logical "0" state in table 1	Input low voltage	V _{IL}
uA	30		V_{DD} can be 1.8V/2.5V/3.3V with $V_{IH} = V_{DD}$	Input high current	I _{IH}
uA		-30	V_{DD} can be 1.8V/2.5V/3.3V with $V_{IH} = V_{DD}$	Input low current	I _{IL}
kΩ		500		Input pullup resistor	R _{pull-} up(IN_SEL)
kΩ		320		Input pulldown resistor	R _{pull-} down(IN_SEL)
		nd 3.3 V ± 5%)	Dlies to $V_{DD} = 1.8 \text{ V} \pm 5\%$, 2.5 V ± 5%	DED LVCMOS/LVTTL CLOCK INPUT (App	SINGLE-END
MHz	250	DC	Clock input	Input frequency	f _{IN}
V	3.465	0.4	Assumes a square wave input with two levels	Single-ended Input Voltage Swing	V _{IN_S-E}
V/ns		0.05		Input Slew Rate (20% to 80% of the amplitude)	dVIN/dt
uA	50		V _{DD} = 3.465 V, V _{IH} = 3.465 V	Input high current	I _{IH}
uA		-30	V _{DD} = 3.465 V, V _{IL} = 0 V	Input low current	I _{IL}
pF		3.5	at 25°C	Input capacitance	C _{IN_SE}
			V ± 5%, 2.5 V ± 5% and 3.3 V ± 5%)	AL CLOCK INPUT (Applies to V _{DD} = 1.8 ^v	DIFFERENTI
GHz	2		Clock input	Input frequency	f _{IN}
,]	2		at 25°C V ± 5%, 2.5 V ± 5% and 3.3 V ± 5%)	Input capacitance AL CLOCK INPUT (Applies to V _{DD} = 1.8 V	I _{IL} C _{IN_SE} DIFFERENTI f _{IN}

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VDD = 1.8 V \pm 5 %, –40 °C \leq T_A \leq 105 °C. Typical values are at VDD = 1.8 V, 25 °C (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
	Differential input voltage peak-to-peak	V _{ICM} = 1 V (V _{DD} = 1.8 V)	0.3		2.4	V_{PP}
$V_{\text{IN,DIFF(p-p)}}$	{2*(V _{INP} -V _{INN})}	V _{ICM} = 1.25 V (V _{DD} = 2.5 V/3.3 V)	0.3		2.4	VPP
V _{ICM}	Input common mode voltage	V _{IN,DIFF(P-P)} > 0.4 V (V _{DD} = 1.8 V/2.5/3.3 V)	0.25		2.3	V
IH	Input high current	V _{DD} = 3.465 V, V _{INP} = 2.4 V, V _{INN} = 1.2 V			30	uA
IL	Input low current	V _{DD} = 3.465 V, V _{INP} = 0 V, V _{INN} = 1.2 V	-30			uA
C _{IN_S-E}	Input capacitance (Single-ended)	at 25°C		3.5		pF
VDS DC OU	JTPUT CHARACTERISTICS					
VOD	Differential output voltage magnitude V _{OUTP} - V _{OUTN}	$V_{IN,DIFF(P-P)} = 0.3 \text{ V}, R_{LOAD} = 100$ Ω	250	350	450	mV
ΔVOD	Change in differential output voltage magnitude. Per output, defined as the difference between VOD in logic hi/lo states.	$V_{IN,DIFF(P-P)} = 0.3 \text{ V, } R_{LOAD} = 100$	–15		15	mV
.,	Steady-state common mode output	$V_{IN,DIFF(P-P)} = 0.3 \text{ V}, R_{LOAD} = 100$ $\Omega \text{ (V}_{DD} = 1.8 \text{ V)}$	1		1.2	V
V _{OC(SS)}	voltage	$V_{IN,DIFF(P-P)} = 0.3 \text{ V}, R_{LOAD} = 100$ $\Omega (V_{DD} = 2.5 \text{ V}/3.3 \text{ V})$	1.1		1.375	V
$\Delta_{VOC(SS)}$	Change in steady-state common mode output voltage. Per output, defined as the difference in VOC in logic hi/lo states.	$V_{\text{IN,DIFF}(P-P)} = 0.3 \text{ V, R}_{\text{LOAD}} = 100$	-15		15	mV
VDS AC OU	JTPUT CHARACTERISTICS					
V_{ring}	Output overshoot and undershoot	$V_{IN,DIFF(P-P)} = 0.3 \text{ V, R}_{LOAD} = 100$ $\Omega, f_{OUT} = 491.52 \text{ MHz}$	-0.1		0.1	V_{OD}
Vos	Output AC common mode	$V_{IN,DIFF(P-P)} = 0.3 \text{ V}, R_{LOAD} = 100$ Ω		50	100	mV_{pp}
os	Short-circuit output current (differential)	V _{OUTP} = V _{OUTN}	-12		12	mA
OS(cm)	Short-circuit output current (common-mode)	V _{OUTP} = V _{OUTN} = 0	-24		24	mA
PD	Propagation delay	$V_{IN,DIFF(P-P)} = 0.3 \text{ V}, R_{LOAD} = 100$ $\Omega^{(1)}$	0.3		0.575	ns
sk, o	Output skew	skew between outputs with the same load conditions (4 and 8 channel) (2)			20	ps
SK, PP	Part-to-part skew	Skew between outputs on different parts subjected to the same operating conditions with the same input and output loading.			250	ps
t _{SK, P}	Pulse skew	50% duty cycle input, crossing point-to-crossing-point distortion (3)	-20		20	ps
RJIT(ADD)	Random additive Jitter (rms)	f_{IN} = 156.25 MHz with 50% duty-cycle, Input slew rate = 1.5V/ns, Integration range = 12kHz - 20MHz, with output load R _{LOAD} = 100 Ω		50	60	fs, RMS
		PN _{1kHz}		-143		
	Phase Noise for a carrier frequency of	PN _{10kHz}		-152		
Phase noise	156.25 MHz with 50% duty-cycle, Input slew rate = 1.5V/ns with output load	PN _{100kHz}		-157		dBc/Hz
		PN _{1MHz}		-160		
		PN _{floor}		-164		

VDD = 1.8 V \pm 5 %, -40 °C \leq T_A \leq 105 °C. Typical values are at VDD = 1.8 V, 25 °C (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
MUX _{ISO}	Mux Isolation	f_{IN} = 156.25 MHz. The difference in power level @ f_{IN} when the selected clock is active and the unselected clock is static versus when the selected clock is inactive and the unselected clock is active.		80		dB
ODC	Output duty cycle	With 50% duty cycle input	45	·	55	%
t _R /t _F	Output rise and fall time	20% to 80% with R_{LOAD} = 100 Ω			300	ps
V _{AC_REF}	Reference output voltage	VDD = 2.5 V, I _{LOAD} = 100 uA	0.9	1.25	1.375	V
POWER SUF	PPLY NOISE REJECTION (PSNR) V _{DD} = 2	.5 V/ 3.3 V				
PSNR	Power Supply Noise Rejection (f _{carrier} =	10 kHz, 100 mVpp ripple injected on V _{DD}		-70		dBc
JUNIX	156.25 MHz)	1 MHz, 100 mVpp ripple injected on V _{DD}		-50		abc

- Measured between single-ended/differential input crossing point to the differential output crossing point.
- (2) (3)
- For the dual bank devices, the inputs are phase aligned and have 50% duty cycle.

 Defined as the magnitude of the time difference between the high-to-low and low-to-high propagation delay times at an output.

6.6 Typical Characteristics

The Figure 6-1 captures the variation of the LMK1D1208 current consumption with input frequency and supply voltage. The LMK1D1204 follows a similar trend. Figure 6-2 shows the variation of the differential output voltage (VOD) swept across frequency. This result is applicable to LMK1D1204 as well.

It is important to note that Figure 6-1 and Figure 6-2 serve as a guidance to the users on what to expect for the range of operating frequency supported by LMK1D120x. It is crucial to note that these graphs were plotted for a limited number of frequencies and load conditions which may not represent the customer system.

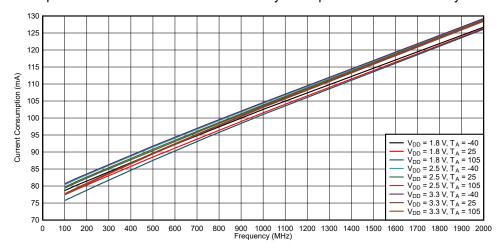


Figure 6-1. LMK1D1208 Current Consumption vs. Frequency

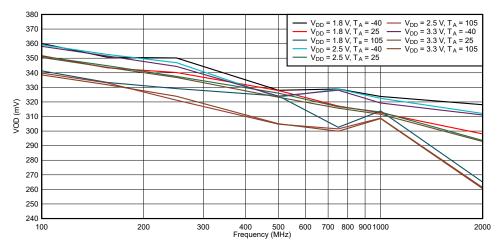


Figure 6-2. LMK1D1208 VOD vs. Frequency



7 Parameter Measurement Information

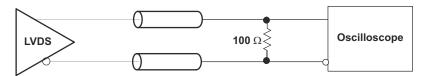


Figure 7-1. LVDS Output DC Configuration During Device Test

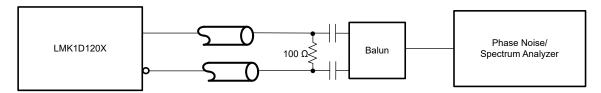


Figure 7-2. LVDS Output AC Configuration During Device Test

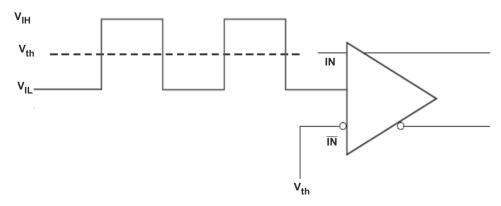


Figure 7-3. DC-Coupled LVCMOS Input During Device Test

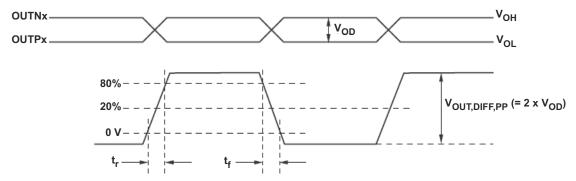
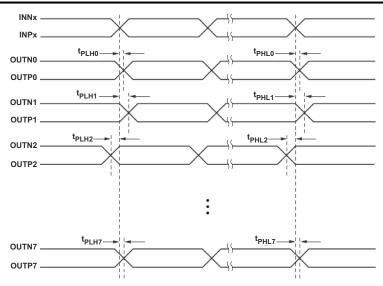


Figure 7-4. Output Voltage and Rise/Fall Time



- A. Output skew is calculated as the greater of the following: the difference between the fastest and the slowest t_{PLHn} or the difference between the fastest and the slowest t_{PHLn} (n = 0, 1, 2, ...7)
- B. Part to part skew is calculated as the greater of the following: the difference between the fastest and the slowest t_{PLHn} or the difference between the fastest and the slowest t_{PHLn} across multiple devices (n = 0, 1, 2, ..7)

Figure 7-5. Output Skew and Part-to-Part Skew

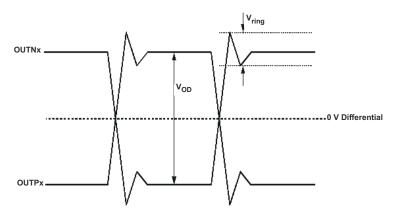


Figure 7-6. Output Overshoot and Undershoot

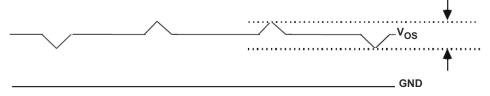


Figure 7-7. Output AC Common Mode

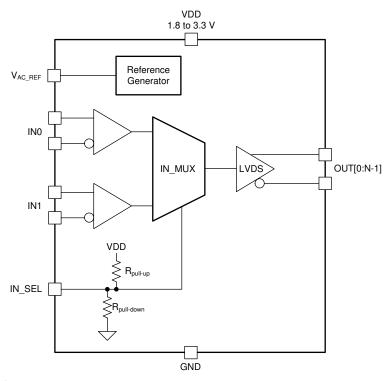
8 Detailed Description

8.1 Overview

The LMK1D120x LVDS drivers use CMOS transistors to control the output current. Therefore, proper biasing and termination are required to ensure correct operation of the device and to maximize signal integrity.

The proper LVDS termination for signal integrity over two $50-\Omega$ lines is $100~\Omega$ between the outputs on the receiver end. Either DC-coupled termination or AC-coupled termination can be used for LVDS outputs. TI recommends placing a termination resistor close to the receiver. If the receiver is internally biased to a voltage different than the output common-mode voltage of the LMK1D12XX, AC-coupling must be used. If the LVDS receiver has internal $100-\Omega$ termination, external termination must be omitted.

8.2 Functional Block Diagram



8.3 Feature Description

The LMK1D120x is a low additive jitter LVDS fan-out buffer that can generate up to eight copies of two selectable LVPECL, LVDS, LP-HCSL, HCSL or LVCMOS inputs. The LMK1D120x can accept reference clock frequencies up to 2 GHz while providing low output skew.

8.3.1 Fail-Safe Input and Hysteresis

The LMK1D120x family of devices is designed to support fail-safe input operation feature. This feature allows the user to drive the device inputs before V_{DD} is applied without damaging the device. Refer to *Specifications* for more information on the maximum input supported by the device. User should note that incorporating the fail-safe inputs also results in a slight increase in clock input pin capacitance.

The device also incorporates an input hysteresis which prevents random oscillation in absence of an input signal. Furthermore, this feature allows the input pins to be left open.

8.3.2 Input Mux

The LMK1D120x family of devices has a 2:1 input mux. This feature allows the user to select between the two clock inputs (using the IN_SEL pin) to the device and fan it out to the outputs. More information on the input selection is provided in the next section.

8.4 Device Functional Modes

The two inputs of the LMK1D120x are internally muxed together and can be selected through the control pin (see Table 8-1). Unused input can be left floating thus reducing the need for additional components. Both AC-and DC-coupling schemes can be used with the LMK1D120x to provide greater system flexibility.

Table 8-1. Input Selection Table

IN_SEL	ACTIVE CLOCK INPUT
0	INO_P, INO_N
1	IN1_P, IN1_N
Open	None (1)

The input buffers are disabled and the outputs are static logic low.

8.4.1 LVDS Output Termination

TI recommends unused outputs to be terminated differentially with a $100-\Omega$ resistor for optimum performance, although unterminated outputs are also okay but will result in slight degradation in performance (Output AC common-mode V_{OS}) in the outputs being used.

The LMK1D120x can be connected to LVDS receiver inputs with DC- and AC-coupling as shown in Figure 8-1 and Figure 8-2 (respectively).

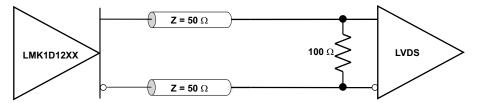


Figure 8-1. Output DC Termination

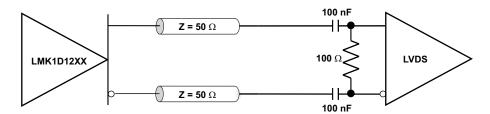


Figure 8-2. Output AC Termination (With the Receiver Internally Biased)

8.4.2 Input Termination

The LMK1D120x input stage is designed with flexibility in mind to allow the user to drive the device with a wide variety of signal types. This device can be interfaced with LVDS, LVPECL, LP-HCSL, HCSL, CML or LVCMOS drivers. Please refer to *Electrical Characteristics* for more details.

LVDS drivers can be connected to LMK1D120x inputs with DC- and AC-coupling as shown Figure 8-3 and Figure 8-4 (respectively).

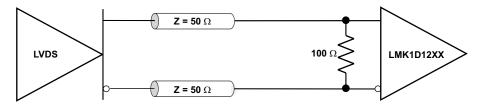


Figure 8-3. LVDS Clock Driver Connected to LMK1D120x Input (DC-Coupled)



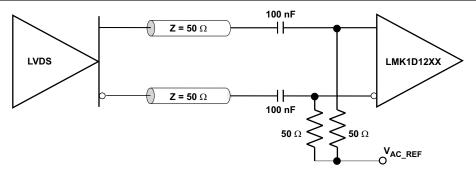


Figure 8-4. LVDS Clock Driver Connected to LMK1D120x Input (AC-Coupled)

Figure 8-5 shows how to connect LVPECL inputs to the LMK1D120x. The series resistors are required to reduce the LVPECL signal swing if the signal swing is $>1.6 \text{ V}_{PP}$.

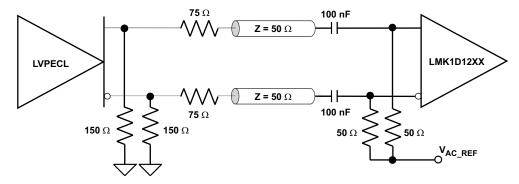


Figure 8-5. LVPECL Clock Driver Connected to LMK1D120x Input

Figure 8-6 illustrates how to couple a LVCMOS clock input to the LMK1D120x directly.

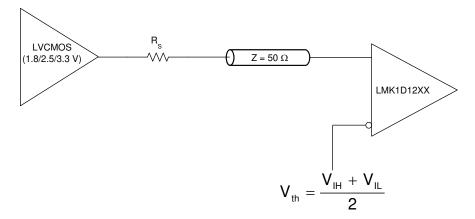


Figure 8-6. 1.8-V/2.5-V/3.3-V LVCMOS Clock Driver Connected to LMK1D120x Input

For unused input, TI recommends grounding both input pins (INP, INN) using 1-k Ω resistors.



9 Application and Implementation

Note

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes, as well as validating and testing their design implementation to confirm system functionality.

9.1 Application Information

The LMK1D120x is a low additive jitter universal to LVDS fan-out buffer with 2 selectable inputs. The small package, low output skew, and low additive jitter make for a flexible device in demanding applications.

9.2 Typical Application

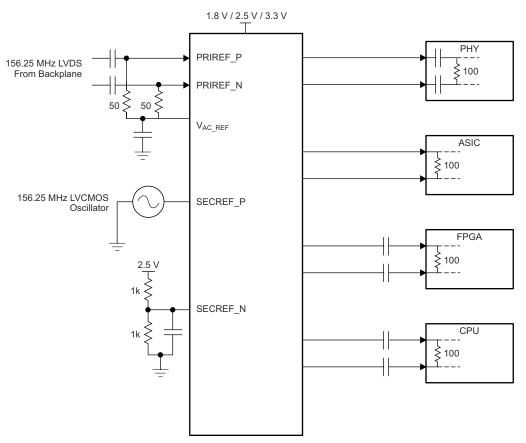


Figure 9-1. Fan-Out Buffer for Line Card Application



9.2.1 Design Requirements

The LMK1D120x shown in Figure 9-1 is configured to select two inputs: a 156.25-MHz LVDS clock from the backplane, or a secondary 156.25-MHz LVCMOS 2.5-V oscillator. The LVDS clock is AC-coupled and biased using the integrated reference voltage generator. A resistor divider is used to set the threshold voltage correctly for the LVCMOS clock. 0.1-µF capacitors are used to reduce noise on both V_{AC_REF} and SECREF_N. Either input signal can be then fanned out to desired devices, as shown. The configuration example is driving 4 LVDS receivers in a line card application with the following properties:

- The PHY device is capable of DC-coupling with an LVDS driver such as the LMK1D120x. This PHY device features internal termination so no additional components are required for proper operation.
- The ASIC LVDS receiver features internal termination and operates at the same common-mode voltage as the LMK1D120x. Again, no additional components are required.
- The FPGA requires external AC-coupling, but has internal termination. 0.1-µF capacitors are placed to provide AC-coupling. Similarly, the CPU is internally terminated, and requires only external AC-coupling capacitors.
- Unused outputs of the LMK1D device are terminated differentially with a 100-Ω resistor for optimum performance.

9.2.2 Detailed Design Procedure

See *Input Termination* for proper input terminations, dependent on single-ended or differential inputs.

See LVDS Output Termination for output termination schemes depending on the receiver application.

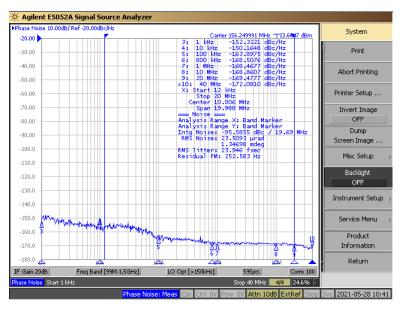
TI recommends unused outputs to be terminated differentially with a $100-\Omega$ resistor for optimum performance, although unterminated outputs are also okay but will result in slight degradation in performance (Output AC common-mode V_{OS}) in the outputs being used.

In this example, the PHY, ASIC, and FPGA or CPU require different schemes. Power-supply filtering and bypassing is critical for low-noise applications.

See *Power Supply Recommendations* for recommended filtering techniques. A reference layout is provided in *Low-Additive Jitter, Four LVDS Outputs Clock Buffer Evaluation Board* (SCAU043).

9.2.3 Application Curves

The LMK1D1208's low additive noise is shown below. The low noise 156.25-MHz source with 24-fs RMS jitter shown in Figure 9-2 drives the LMK1D1208, resulting in 46.4-fs RMS when integrated from 12 kHz to 20 MHz (Figure 9-3). The resultant additive jitter is a low 39.7-fs RMS for this configuration. Note that this result applies to the LMK1D1204 device as well.



A. Reference signal is low-noise Rohde and Schwarz SMA100B

Figure 9-2. LMK1D208 Reference Phase Noise, 156.25 MHz, 24-fs RMS (12 kHz to 20 MHz)

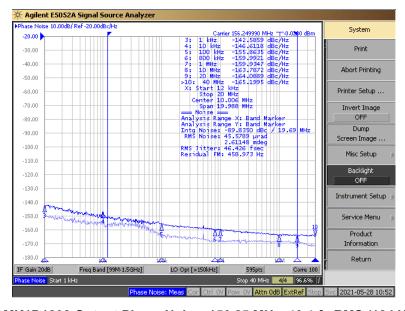


Figure 9-3. LMK1D1208 Output Phase Noise, 156.25 MHz, 46.4-fs RMS (12 kHz to 20 MHz)



The Figure 9-4 captures the low close-in phase noise of the LMK1D1208 device. The LMK1D1204 and LMK1D1208 have excellent flicker noise as a result of superior process technology and design. This enables their use for clock distribution in radar systems, medical imaging systems etc which require ultra-low close-in phase noise clocks.

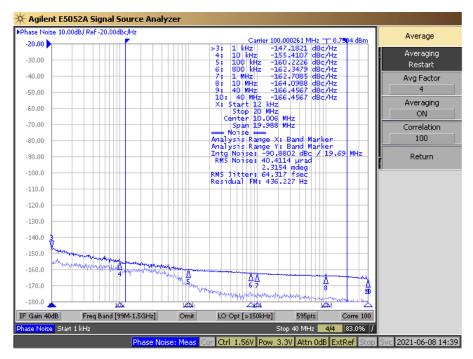


Figure 9-4. LMK1D1208 Output Phase Noise, 100 MHz, 1 kHz offset: -147 dBc/Hz

10 Power Supply Recommendations

High-performance clock buffers are sensitive to noise on the power supply, which can dramatically increase the additive jitter of the buffer. Thus, it is essential to reduce noise from the system power supply, especially when jitter or phase noise is critical to applications.

Filter capacitors are used to eliminate the low-frequency noise from the power supply, where the bypass capacitors provide the low impedance path for high-frequency noise and guard the power-supply system against the induced fluctuations. These bypass capacitors also provide instantaneous current surges as required by the device and must have low equivalent series resistance (ESR). To properly use the bypass capacitors, they must be placed close to the power-supply pins and laid out with short loops to minimize inductance. TI recommends adding as many high-frequency (for example, 0.1-µF) bypass capacitors as there are supply pins in the package. TI recommends, but does not require, inserting a ferrite bead between the board power supply and the chip power supply that isolates the high-frequency switching noises generated by the clock driver; these beads prevent the switching noise from leaking into the board supply. Choose an appropriate ferrite bead with low DC-resistance because it is imperative to provide adequate isolation between the board supply and the chip supply, as well as to maintain a voltage at the supply pins that is greater than the minimum voltage required for proper operation.

Figure 10-1 shows this recommended power-supply decoupling method.

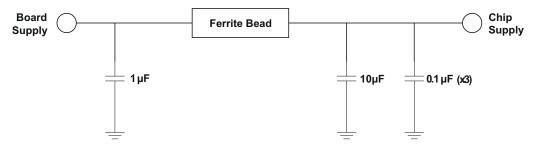


Figure 10-1. Power Supply Decoupling



11 Layout

11.1 Layout Guidelines

For reliability and performance reasons, the die temperature must be limited to a maximum of 135°C.

The device package has an exposed pad that provides the primary heat removal path to the printed circuit board (PCB). To maximize the heat dissipation from the package, a thermal landing pattern including multiple vias to a ground plane must be incorporated into the PCB within the footprint of the package. The thermal pad must be soldered down to ensure adequate heat conduction to of the package. Figure 11-1 shows a recommended land and via pattern for LMK1D1208.

11.2 Layout Example

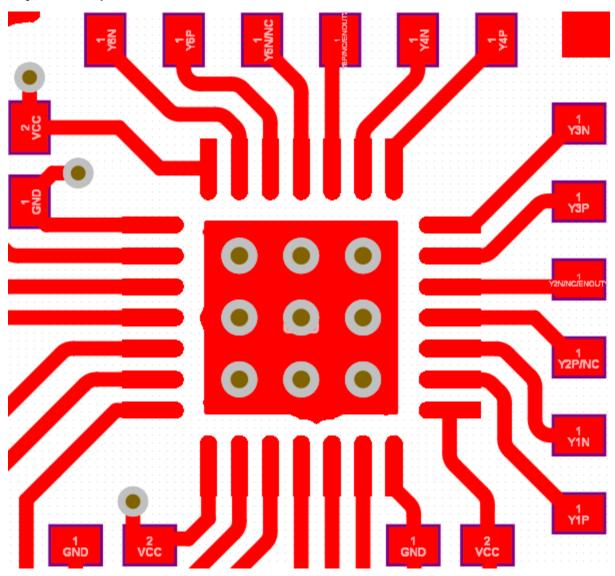


Figure 11-1. Recommended PCB Layout, Top Layer



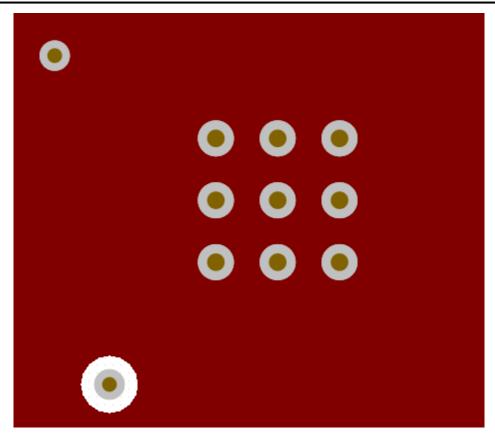


Figure 11-2. PCB Layout, GND Layer



12 Device and Documentation Support

12.1 Documentation Support

12.1.1 Related Documentation

For related documentation see the following:

- Low-Additive Jitter, Four LVDS Outputs Clock Buffer Evaluation Board (SCAU043)
- Power Consumption of LVPECL and LVDS (SLYT127)
- Semiconductor and IC Package Thermal Metrics (SPRA953)
- Using Thermal Calculation Tools for Analog Components (SLUA556)

12.2 Receiving Notification of Documentation Updates

To receive notification of documentation updates, navigate to the device product folder on ti.com. Click on *Subscribe to updates* to register and receive a weekly digest of any product information that has changed. For change details, review the revision history included in any revised document.

12.3 Support Resources

TI E2E[™] support forums are an engineer's go-to source for fast, verified answers and design help — straight from the experts. Search existing answers or ask your own question to get the quick design help you need.

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12.5 Electrostatic Discharge Caution



This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

12.6 Glossary

TI Glossary

This glossary lists and explains terms, acronyms, and definitions.

13 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

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PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan	Lead finish/ Ball material	MSL Peak Temp	Op Temp (°C)	Device Marking (4/5)	Samples
LMK1D1204RGTR	ACTIVE	VQFN	RGT	16	3000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 105	LD1204	Samples
LMK1D1204RGTT	ACTIVE	VQFN	RGT	16	250	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 105	LD1204	Samples
LMK1D1208RHDR	ACTIVE	VQFN	RHD	28	3000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 105	LMK1D 1208	Samples
LMK1D1208RHDT	ACTIVE	VQFN	RHD	28	250	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 105	LMK1D 1208	Samples

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

(2) RoHS: TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

Green: TI defines "Green" to mean the content of Chlorine (CI) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

- (3) MSL, Peak Temp. The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.
- (4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.
- (5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.
- (6) Lead finish/Ball material Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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PACKAGE OPTION ADDENDUM

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continues to take reasonable steps to provide representative and accurate information but may not have conducted destructive testing or chemical analysis on incoming materials and chemicals. TI and TI suppliers consider certain information to be proprietary, and thus CAS numbers and other limited information may not be available for release.

In no event shall TI's liability arising out of such information exceed the total purchase price of the TI part(s) at issue in this document sold by TI to Customer on an annual basis.

PACKAGE MATERIALS INFORMATION

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TAPE AND REEL INFORMATION





A0	Dimension designed to accommodate the component width
В0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



*All dimensions are nominal

Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
LMK1D1204RGTR	VQFN	RGT	16	3000	330.0	12.4	3.3	3.3	1.1	8.0	12.0	Q2
LMK1D1204RGTT	VQFN	RGT	16	250	180.0	12.4	3.3	3.3	1.1	8.0	12.0	Q2
LMK1D1208RHDR	VQFN	RHD	28	3000	330.0	12.4	5.3	5.3	1.1	8.0	12.0	Q2
LMK1D1208RHDT	VQFN	RHD	28	250	180.0	12.4	5.3	5.3	1.1	8.0	12.0	Q2

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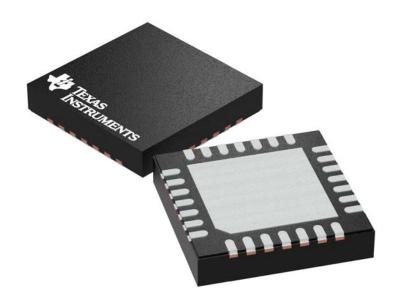


*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
LMK1D1204RGTR	VQFN	RGT	16	3000	367.0	367.0	35.0
LMK1D1204RGTT	VQFN	RGT	16	250	210.0	185.0	35.0
LMK1D1208RHDR	VQFN	RHD	28	3000	367.0	367.0	35.0
LMK1D1208RHDT	VQFN	RHD	28	250	210.0	185.0	35.0

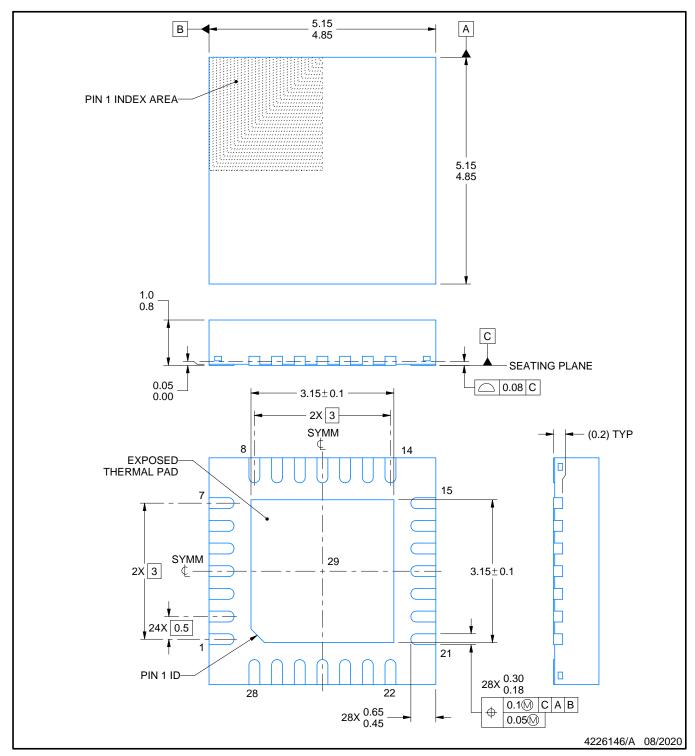
5 x 5 mm, 0.5 mm pitch

PLASTIC QUAD FLATPACK - NO LEAD



Images above are just a representation of the package family, actual package may vary. Refer to the product data sheet for package details.

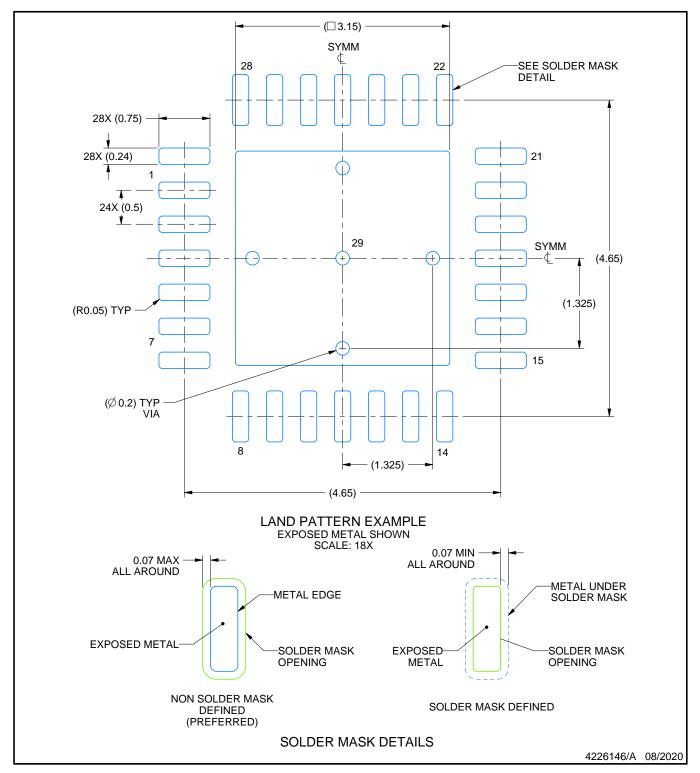




NOTES:

- 1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
 2. This drawing is subject to change without notice.
- 3. The package thermal pad must be soldered to the printed circuit board for thermal and mechanical performance.

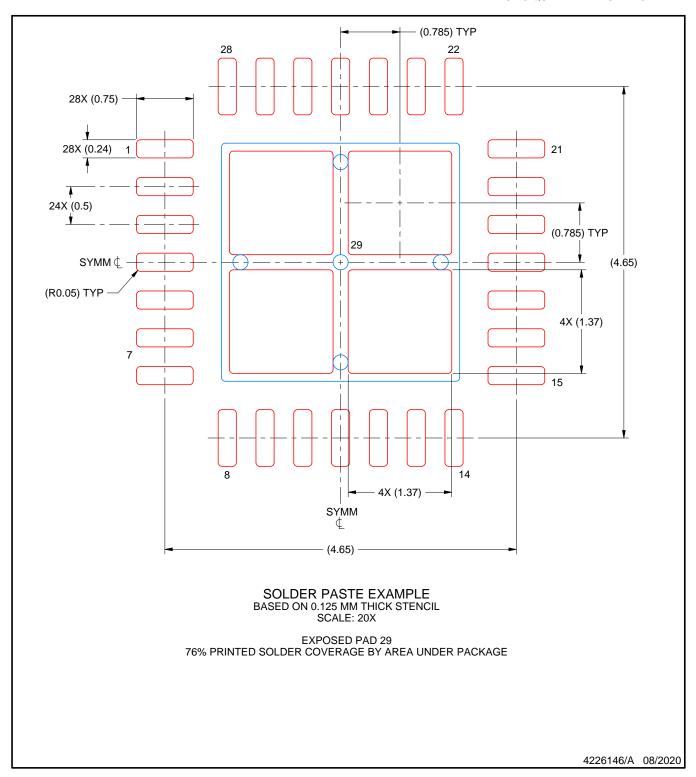




NOTES: (continued)

- 4. This package is designed to be soldered to a thermal pad on the board. For more information, see Texas Instruments literature number SLUA271 (www.ti.com/lit/slua271).
- 5. Vias are optional depending on application, refer to device data sheet. If any vias are implemented, refer to their locations shown on this view. It is recommended that vias under paste be filled, plugged or tented.





NOTES: (continued)

^{6.} Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.



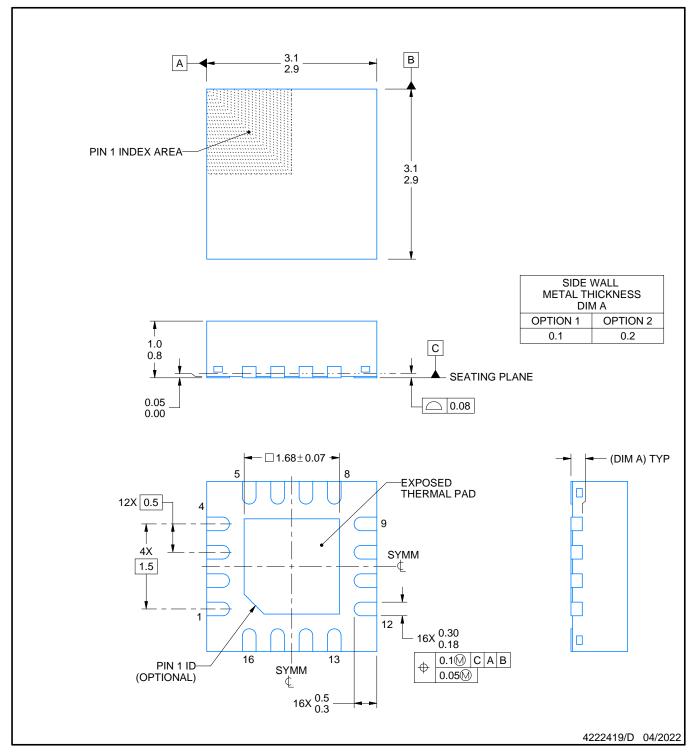


Images above are just a representation of the package family, actual package may vary. Refer to the product data sheet for package details.





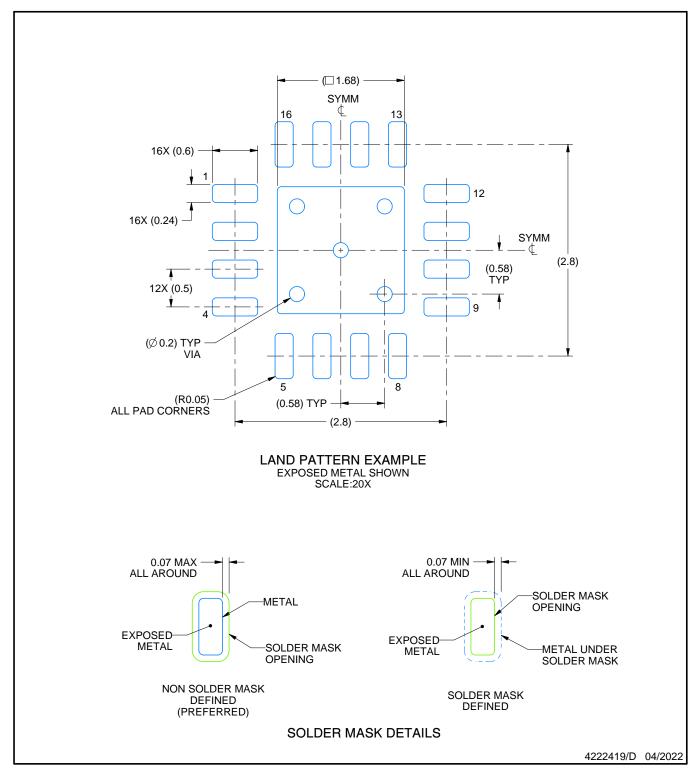




NOTES:

- 1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
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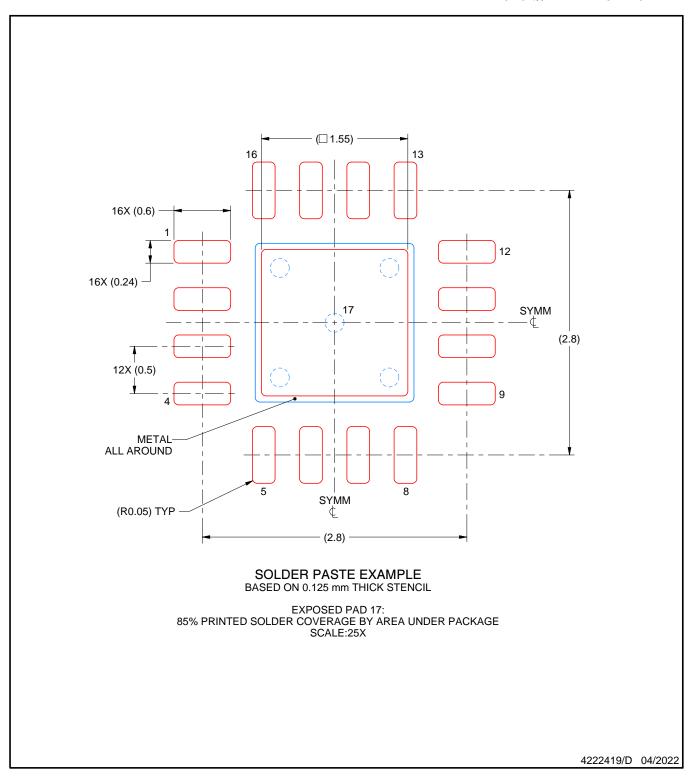




NOTES: (continued)

- 4. This package is designed to be soldered to a thermal pad on the board. For more information, see Texas Instruments literature number SLUA271 (www.ti.com/lit/slua271).
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NOTES: (continued)

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